

# FM Spectroscopy

Or, How Excessive Trigonometry Solves All Problems

David Grabovsky

Columbia University

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## Laser Locking for 5-Year-Olds

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How do you make it stay put?*

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- 1 Create a signal with two frequency sidebands around the central laser frequency; send this through an atomic sample.
- 2 Interfere the sidebands with the transmitted beam; the beats contain absorption and dispersion data as an AM signal!
- 3 Heterodyne with the sidebands to get an error signal.
- 4 Construct a feedback loop to keep the laser locked!

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- 1 Basic FM Theory
- 2 Phase Modulation
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# Modulating a Signal

- Consider a signal of the form  $s(t) = A_c \cos(\omega_0 t + \phi)$ .
- Its frequency is constant, so the phase rises as  $\omega_0 t + \phi$ .
- If we “modulate” the signal by  $\omega_0 \mapsto \omega(t) = \omega_0 + m(t)$ , this is no longer true: the phase must be realized as an integral.

The modulated signal is therefore written

$$\begin{aligned} s_m(t) &= A_c \cos \left( \int_0^t \omega(\tau) d\tau \right) = \\ &= A_c \cos \left( \omega_0 t + \int_0^t m(\tau) d\tau \right). \end{aligned} \quad (1.1)$$

## Sinusoidal Modulation

Suppose that we modulate sinusoidally:  $m(t) = A_m \cos(\Omega t)$ .  
 Its integral is  $\int_0^t m(\tau) d\tau = \frac{A_m}{\Omega} \sin(\Omega t)$ , and so

$$s_m(t) = A_0 \cos \left[ \omega_0 t + \frac{\Delta\omega}{\Omega} \sin(\Omega t) \right], \quad (1.2)$$

where  $\Delta\omega \equiv A_m$  is called the **peak frequency deviation** and is described by the amplitude—*not* the frequency—of  $m(t)$ .

*This is Chekhov's gun, the loaded pistol gingerly laid down on the table in Act I of a drama. Commit it to mind; it will return.*



# Bessel Functions

For the modulated signal  $s_m(t) = A_0 \cos \left[ \omega_0 t + \frac{\Delta\omega}{\Omega} \sin(\Omega t) \right]$ ,  
define the **modulation index**  $M \equiv \frac{\Delta\omega}{\Omega}$ .

## FACT OF LIFE.

A sinusoidally modulated signal can be expressed in terms of the Bessel functions:

$$s_m(t) = \cos \left[ \omega_0 t + \frac{\Delta\omega}{\Omega} \sin(\Omega t) \right] = \sum_{n=-\infty}^{\infty} J_n(M) \cos [(\omega_0 + n\Omega)t]. \quad (1.3)$$

## Bessel Functions: Discussion

We now return to physics; i.e. making ridiculous approximations and claiming that asymptotics make everything obvious.

- In general,  $s_m(t)$  has infinite bandwidth, but fortunately the  $J_n$  decay quickly:  $J_n(M) \approx \frac{M^n}{2^n n!}$ , so wide sidebands are small.
- We care about the regime  $M \ll 1$  (“narrowband FM”), where  $\Delta\omega \ll \Omega$ , and the “bandwidth” is about  $2\Omega$ . Indeed...
- For  $M \ll 1$ , the asymptotic  $J_n \approx \frac{M^n}{2^n n!}$  permits higher powers  $M^k$  to vanish if we chant “to first order” and wave our hands.
- Moreover, we obtain  $J_0(M) \approx 1$ ,  $J_{\pm 1}(M) = \pm \frac{M}{2}$ , and all of the other Bessel functions are banished to Siberia.

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## The Basic Setup

The laser beam provides an electric field

$$\mathbf{E}(t) = \mathbf{E}_0 e^{i\omega_0 t} + c.c.; \quad (2.1)$$

this is promptly modulated as in (1.2). Its frequency now oscillating sinusoidally, the field careers towards the sample:

$$\mathbf{E}_m(t) = \mathbf{E}_0 \exp [i\omega_0 t + iM \sin(\Omega t)] + c.c., \quad (2.2)$$

where  $M$  is the modulation index. In general, as per (1.3),

$$\mathbf{E}_m(t) = \mathbf{E}_0 \sum_{n=-\infty}^{\infty} J_n(M) \exp [i(\omega_0 + n\Omega)t] + c.c. \quad (2.3)$$

But we are not savages, and for  $M \ll 1$  there's a neat trick.

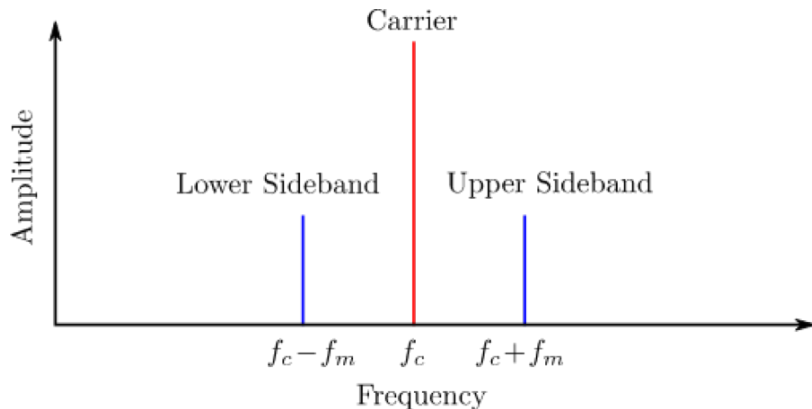
# A Smol Approximation

We approximate  $E(t)$  to first order in  $M$ :

$$\begin{aligned} E(t) &\sim \exp [i\omega_0 t + iM \sin(\Omega t)] = \\ &= e^{i\omega_0 t} \exp \left[ \frac{M}{2} (e^{i\Omega t} - e^{-i\Omega t}) \right] \approx \\ &\approx e^{i\omega_0 t} \left[ 1 + \frac{M}{2} (e^{i\Omega t} - e^{-i\Omega t}) \right] \implies \end{aligned}$$

$$\boxed{\mathbf{E}(t) = \mathbf{E}_0 \left\{ e^{i\omega_0 t} + \frac{M}{2} [e^{i(\omega_0+\Omega)t} - e^{i(\omega_0-\Omega)t}] \right\}. \quad (2.4)}$$

## The Result: Carrier and Sidebands



**Figure:** Small sidebands of amplitude  $M/2$  around a carrier frequency. It will be our goal to lock this central frequency to an atomic resonance.

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## Transmitted Field

- $\mathbf{E}(t)$  sojourns through an atomic sample with a known resonance.
- The amounts by which its amplitudes and phases are thrown off are functions  $\alpha(\omega), \phi(\omega)$  of frequency, but each frequency in the signal is distorted independently.
- A photodiode then detects the intensity of the transmitted field, which scales as  $V_T(t) \sim |E_T(t)|^2$ . The 3 terms of  $E_T$  will interfere, and we will keep terms to first order in  $M$ .

Take a deep breath...



## Excessive Trigonometry

The transmitted field is

$$E_T(t) \sim \alpha(\omega_0) e^{i\omega_0 t} e^{i\phi(\omega_0)} + \frac{M}{2} \left[ \alpha(\omega_0 + \Omega) e^{i(\omega_0 + \Omega)t} e^{i\phi(\omega_0 + \Omega)} - \alpha(\omega_0 - \Omega) e^{i(\omega_0 - \Omega)t} e^{i\phi(\omega_0 - \Omega)} \right]. \quad (3.1)$$

To first order in  $M$ , the photodiode detects

$$V_T(t) \sim \frac{M}{2} \alpha(\omega_0) \left[ 1 \pm \alpha(\omega_0 \pm \Omega) \left( e^{i\Omega t} e^{\mp i(\phi(\omega_0) - \phi(\omega_0 \pm \Omega))} + c.c. \right) \right]. \quad (3.2)$$

## Excessive Trigonometry

...However, we can rewrite this by filling seven pages with sines and cosines and my head with existential despair at 4am. The result is:

$$V_T(t) \sim \alpha(\omega_0) \left[ 1 + \frac{M}{2} (\Delta\alpha \cos(\Omega t) + \Delta\phi \sin(\Omega t)) \right], \quad (3.3)$$

where we have introduced

$$\Delta\alpha \equiv \alpha(\omega_0 - \Omega) - \alpha(\omega_0 + \Omega); \quad (3.4)$$

$$\begin{aligned} \Delta\phi &\equiv \phi(\omega_0 + \Omega) + \phi(\omega_0 - \Omega) - 2\phi(\omega_0) = \\ &= 2 \left[ \bar{\phi}(\omega_0 \pm \Omega) - \phi(\omega_0) \right]. \end{aligned} \quad (3.5)$$

## What's Going On Here?

- Computing  $V_T(t)$  amounts to interfering the sidebands with the transmitted carrier signal. The signal now has an interference envelope (with beat frequency  $\Omega$ ) whose *amplitude* encodes the frequency-dependent absorption and dispersion functions.
- The in-phase component measures the absorption difference between the sidebands, and the quadrature component measures the difference between the average of the sidebands' phase shift and the carrier's phase shift.
- Also, if  $\Omega$  is small, then the cosine term becomes proportional to the derivative of the absorption, while the sine term becomes proportional to the second derivative of the dispersion.

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# Mixing Signals

*If only there were a way to extract  $\Delta\alpha$  and  $\Delta\phi$ ...*

Recall the trigonometric identities

$$\sin(\theta_1 t) \sin(\theta_2 t) = \frac{1}{2} [\cos((\theta_1 - \theta_2)t) - \cos((\theta_1 + \theta_2)t)]; \quad (4.1)$$

$$\sin(\theta_1 t) \cos(\theta_2 t) = \frac{1}{2} [\sin((\theta_1 - \theta_2)t) + \sin((\theta_1 + \theta_2)t)]. \quad (4.2)$$

- If we “mix” two sinusoidal signals by multiplying them, we can then use a low-pass filter to isolate the beat frequency between them.
- Suppose we mixed  $V_T(t)$  with an internal-oscillator signal  $\sin(\Omega t + \theta)$ , where  $\theta$  is a tunable phase.

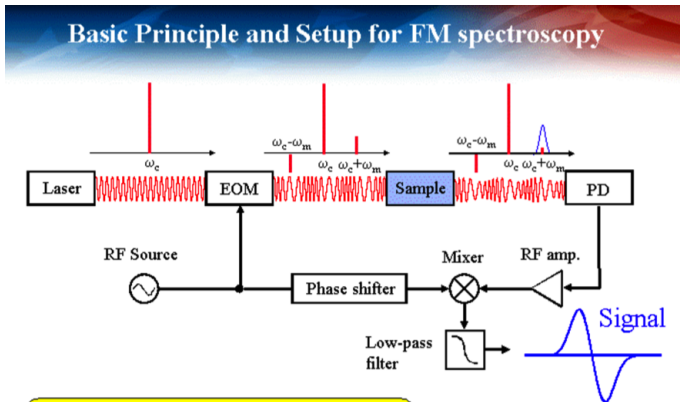
## Locking a Laser

Applying a low-pass filter to the mixed signal, we would recover

$$\Delta\alpha \sin(\theta) + \Delta\phi \cos(\theta). \quad (4.3)$$

Since  $\theta$  is tunable, we can crank it up to  $\pi$  to study absorption, or down to 0 to study dispersion. And indeed, both  $\Delta\alpha$  and  $\Delta\theta$  have sharp zero crossings when the resonance condition is achieved.

# High-Level Diagram



Simultaneous comparison between on-resonant and off-resonant cases

# Gaussian Spectral Feature

